

Voltage and Frequency Control of Isolated Hybrid Power Systems

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Abstract: — This paper proposes A hybrid VFDC/PQDC-based MG control scheme for Microgrid(MG) applications particularly islanded applications. A microgrid can be operated in both grid-connected and in autonomous mode. The control of autonomous MG is very complex without the support of public utility and grid. The control of virtual impedance plays an important role to decouple the active and reactive power allocations for energy storage systems (ESS). In this paper, ESSs control method is proposed with V/f droop control (VFDC) and P/Q droop control (PQDC) combined. The proposed method results are verified by using MATLAB SIMULINK.

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I. INTRODUCTION

Generally a Microgrid (MG) has two modes of operations, they are a) autonomous operation mode; and b) grid-connected operation mode [1]–[3]. The control of an autonomous operation mode is very complex comparatively grid connected operation mode at where support of main grid is absent. Since inertia of MG is in smaller in size comparatively public grid, so the medium and small interruptions particularly the power fluctuations in solar arrays and wind turbines and also in load cut-off and load cut-in may lead to power quality issues. In these conditions to control the system voltage and frequency reliable power sources are required.

Energy storage system (ESS) can be controlled to release or absorb active/reactive power flexibly. With the appropriate control of active/reactive power by ESS, frequency and voltage can be maintained stable. [4]–[6]. constant voltage and frequency control (V/f) can be get by a single MPS for a medium or small size MG, which control ESS as a constant voltage/frequency source. For a large size MG two or more ESSs are required, which increases the system control complexity. V/f control method is usually invalid due to the loop current among the parallel connected ESSs.

Droop control is another method used to control the voltage and frequency, which imitates the characteristics of synchronous machine on the transmission level[7]. Based on the out-put active power and reactive power, the voltage angle and amplitude of each ESS is regulated respectively. Therefore, the active power and reactive power can be shared according the preset droop coefficient among the parallel ESSs, and the cut-in or cut-off of any of the ESS unit affects little to MG normal operation. Droop control makes power sources “plug and play” possible, which is an important feature for advanced MG [1].

Traditional droop control method is developed for high voltage (HV) transmission system in which the line impedance is very high. In MG the line impedance is resistive and the voltage levels are low/medium. So the traditional droop control method is not valid for MG due to active and reactive power coupling among ESSs, which will decrease the system transient response and steady-state performance [8]. In this situation due to the uncertainty of line impedance and control parameters the control process becomes more complex [9]. To make the power lines to be more inductive virtual impedance is used at converters [10]–[13].

This paper proposes a novel control method for ESSs-based MG which combines traditional V/f droop control (VFDC) with P/Q droop control (PQDC). The VFDC ESS provides the frequency and voltage determined by the actual active power and reactive power, while the PQDC ESS provides certain active power and reactive power determined by system frequency and voltage deviations from their nominal values. Since PQDC is independent of line parameters and other control errors, the interaction among ESSs can be avoided and system stability is improved. In this way, the frequency and voltage can be maintained stable through proper active power and reactive power distribution among all the parallel connected ESSs.

The rest of this paper is organized as follows. Section II describes the features of MPSs with typical control methods for autonomous MG. Section III presents the proposed hybrid VFDC/PQDC control scheme. Section IV gives the experimental results. Finally, Section V presents the conclusion of this paper.

II. CONTROL OF MPS FOR AUTONOMOUS MG

As there is no synchronous generator available in MG, establishing and maintaining system frequency and voltage i.e., the choice and control of power supply is a big problem in autonomous MG operation. ESS with converter can be controlled flexibly with different output characteristics and is considered as a potential MPS for MG autonomous operation [21]. (For an autonomous MG, ESS can be considered as MPS when ESS placed with a converter)

Generally there are two methods to control ESS as a MPS a) V/f control b) droop control. Compared to V/f control, current sharing effect is good in droop control. So this paper focuses on droop control. Traditional droop control named in this paper as VFDC is analyzed and compared with the proposed hybrid VFDC/PQDC.

A. VFDC

Traditional synchronous characteristics are mimics by VFDC-based ESS with voltage angle and amplitude regulated according to its output active power and reactive power.

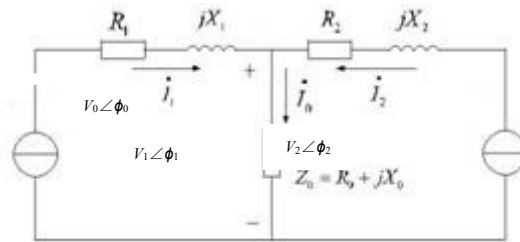


Fig. 1. MG equivalent circuit based on VFDC method.

The equivalent circuit of a MG with VFDC-based ESSs shown in Fig. 1, where the two ESSs are equivalent as voltage sources $V_1\angle\phi_1$ and $V_2\angle\phi_2$, respectively.

The load voltage ($\bar{V}_0$) is derived as
The load voltage ($\bar{V}_0$) is derived as

$$\bar{I}_0 = \bar{I}_1 + \bar{I}_2$$

$$\text{So, } \bar{V}_0 = \frac{\frac{\bar{V}_1 + \bar{V}_2}{\frac{1}{Z_1} + \frac{1}{Z_2}}}{\frac{1}{Z_0} + \frac{1}{Z_1} + \frac{1}{Z_2}} \quad (1)$$

where $Z_1 = R_1 + jX_1$ and $Z_2 = R_2 + jX_2$ are equivalent line impedance of $V_1\angle\phi_1$ and $V_2\angle\phi_2$, respectively. The output current of the two voltage sources are

$$\bar{I}_1 = \frac{\bar{V}_1 - \bar{V}_0}{Z_1} = \frac{(\bar{Z}_2 + \bar{Z}_0)\bar{V}_1 - \bar{Z}_0\bar{V}_2}{Z_1 Z_2 + Z_0 Z_2 + Z_0 Z_1} \quad (2)$$

$$\bar{I}_2 = \frac{\bar{V}_2 - \bar{V}_0}{Z_2} = \frac{(\bar{Z}_1 + \bar{Z}_0)\bar{V}_2 - \bar{Z}_0\bar{V}_1}{Z_1 Z_2 + Z_0 Z_2 + Z_0 Z_1} \quad (3)$$

The apparent power of the two ESSs

$$\bar{S}_1 = \bar{V}_1 \times \bar{I}_1^* = \bar{V}_1^* \times \frac{(\bar{Z}_2^* + \bar{Z}_0^*)\bar{V}_1^* - \bar{Z}_0^*\bar{V}_2^*}{Z_1^* Z_2^* + Z_0^* Z_2^* + Z_0^* Z_1^*} \quad (4)$$

$$\bar{S}_2 = \bar{V}_2 \times \bar{I}_2^* = \bar{V}_2^* \times \frac{(\bar{Z}_1^* + \bar{Z}_0^*)\bar{V}_2^* - \bar{Z}_0^*\bar{V}_1^*}{Z_1^* Z_2^* + Z_0^* Z_2^* + Z_0^* Z_1^*} \quad (5)$$

where the superscript * is the corresponding conjugate phasors.

From the above eq.(2)-(5), it is clear that, the output current and apparent power of the two ESSs are dependent on the source voltage values ($\bar{V}_1$, $\bar{V}_2$) and line impedances (Z_1 , Z_2).

The current sharing of ESSs can only be implemented by the coordinated control of $\bar{V}_1$ and $\bar{V}_2$

$$\bar{I}_1 = \bar{I}_2$$

$$\frac{(\bar{Z}_2 + \bar{Z}_0)\bar{V}_1 - \bar{Z}_0\bar{V}_2}{Z_1 Z_2 + Z_0 Z_2 + Z_0 Z_1} = \frac{(\bar{Z}_1 + \bar{Z}_0)\bar{V}_2 - \bar{Z}_0\bar{V}_1}{Z_1 Z_2 + Z_0 Z_2 + Z_0 Z_1} \quad (6)$$

The phase angle difference between $\bar{V}_1$ and $\bar{V}_2$ is

$$\angle(\theta_2 - \theta_1) = \frac{V_2 (\bar{Z}_1 + 2\bar{Z}_0)}{V_1 \bar{Z}_2 + 2\bar{Z}_0} \quad (7)$$

Where θ_1 and θ_2 , V_1 and V_2 are the phase angle and voltage amplitude of $\bar{V}_1$ and $\bar{V}_2$ respectively.

$$\cos(\theta_1 - \theta_2) = \frac{V_2 (2R_0 + R_1)(2R_0 + R_2) + (2X_0 + X_1)(2X_0 + X_2)}{V_1 ((2R_0 + R_2)^2 + (2X_0 + X_2)^2)} \quad (8)$$

$$\sin(\theta_1 - \theta_2) = \frac{V_2 (2X_0 + X_1)(2R_0 + R_2) - (2R_0 + R_1)(2X_0 + X_2)}{V_1 ((2R_0 + R_2)^2 + (2X_0 + X_2)^2)} \quad (9)$$

From eq (8) and (9) it is clear that the power distribution in the system is affected by line impedances, so it increases the complexity in control of ESS and current sharing effect reduces due to the uncertainty in line parameters.

B. Hybrid VFDC/PQDC

PQDC-based ESS controls its output active power and reactive power based on MG frequency and voltage changes from their rated values. Let the ESS V_2 is regulated by PQDC, then the equivalent circuit shown in fig.1 can be described as a connection of $V_1 \angle \phi_1$ and $I_2 \angle \phi_2$ as shown below.

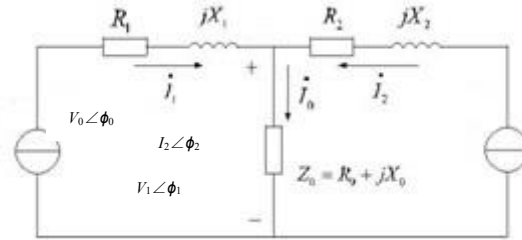


Fig. 2. MG equivalent circuit based on PQDC method.

By applying KCL and KVL, $\bar{I}_1 = \frac{\bar{V}_1 - \bar{I}_2 \bar{Z}_0}{\bar{Z}_0 + \bar{Z}_1}$ (10)

$$\text{And } \bar{S}_1 = \bar{V}_1 \times \bar{I}_1^* = \bar{V}_1^* \times \left(\frac{\bar{V}_1 - \bar{I}_2 \bar{Z}_0}{\bar{Z}_0 + \bar{Z}_1} \right)^* = \frac{V_1^2 - \bar{Z}_0^* \bar{I}_2^* \bar{V}_1}{\bar{Z}_0^* + \bar{Z}_1^*} \quad (11)$$

From eq(11), $\bar{S}_1$ is not depends on Z_2 . If $\bar{V}_1$ and line parameters are given, then $\bar{S}_1$ can found only by the value of $\bar{I}_2$. From the above we can say that, the diversity and uncertainty in line parameters(Z_1 , Z_2) has no impact on MG power regulation. This means that the hybrid VFDC/PQDC system is better in current sharing control than the VFDC system.

III. PROPOSED HYBRID VFDC/PQDC

CONTROL SCHEME

The VFDC-based ESS is controlled as a V/f droop voltage source with output voltage and frequency determined by its active power and reactive power. While the PQDC-based ESS is controlled as a P/Q source with output active power and reactive power determined by the voltage and frequency deviations on its output terminal. To decouple the active power and reactive power due to the resistive line parameters, virtual impedance control is adopted.

A. Virtual Impedance Control

From [22], the real and reactive power allotted to two converters connected to a common bus are given as

$$P = \left(\frac{EV}{Z} \cos \phi - \frac{V^2}{Z} \right) \cos \theta + \frac{EV}{Z} \sin \phi \sin \theta \quad (12)$$

$$Q = \left(\frac{EV}{Z} \cos \phi - \frac{V^2}{Z} \right) \sin \theta - \frac{EV}{Z} \sin \phi \cos \theta \quad (13)$$

where E and V are the amplitudes of converters output voltage and bus voltage, ϕ is the power angle, Z and θ are the magnitude and the phase of output impedance.

For HV transmission system line impedance is more inductive, so θ is assumed as 90° , $Z \approx X$, where X is line reactance.

Then the above eq(12-13) are

$$P = \frac{EV}{X} \sin \phi; \quad Q = \frac{EV \cos \phi - V^2}{X} \quad (14)$$

For low voltage distribution line impedance is more resistive in nature so θ is assumed as 0° , $Z \approx R$, where R is line resistance.

Then the above eq(12-13) are

$$P = \frac{EV \cos \phi - V^2}{R}, \quad Q = -\frac{EV}{R} \sin \phi \quad (15)$$

By observing eq (14) and (15), the real and reactive power distribution between converters is effected by line impedance. low/medium voltage MG has the line impedance as resistive but the P - f and Q - v decoupling of traditional droop control is done on inductive impedance. So to control MG, we have to make the line impedance to be inductive [23], this can be done with the help of Virtual impedance control method, which makes the converter output impedance to be inductive and then the total impedance is inductive [19]

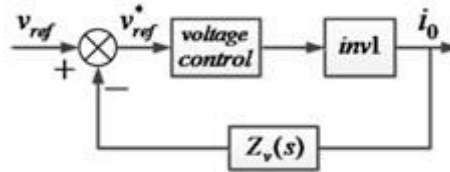


Fig. 3. Schematic of virtual impedance control loop

The schematic virtual impedance control loop is shown in fig.3, where $Z_v(s)$ is the virtual impedance equivalent as an inductor L_v .

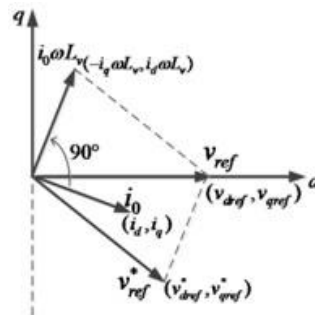


Fig. 4. Phasor diagram of virtual impedance control based on d-q axis.

Truly speaking virtual impedance is a voltage correction method, as shown in fig.4, where v_{ref} is the reference voltage, v_{ref}^* is that after modification, i_0 is the output current, and $i_0 \omega L_v$ is the voltage drop on virtual inductor L_v .

The reference output voltage can be given as,

$$v_{dref}^* = v_{dref} + i_q \omega L_v \quad (16)$$

$$v_{qref}^* = v_{qref} + i_d \omega L_v \quad (17)$$

As a result, the equivalent impedance is controlled to be inductive which enables P - f and Q - v decoupling. Moreover, the active and reactive current sharing among the parallel connected ESSs can be improved since their output impedances are definite.

B. VFDC Control

The VFDC-based ESS is controlled to output certain frequency and voltage according to its active and reactive power. Since the virtual inductor L_v will absorb certain reactive power Q_v , it is taken as a correction to the reactive power reference Q^* . The droop control loop with reactive power compensation is illustrated in Fig. 5.

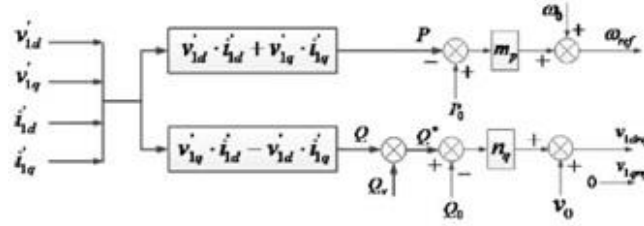


Fig. 5. VFDC control loop with reactive power compensation.

$$\begin{aligned}\omega_{ref} &= \omega_0 + m_p(P_0 - P) \\ v_{ref} &= v_0 + n_q(Q_0^* - Q)\end{aligned}\quad (17)$$

Where, ω_{ref} , v_{ref} are the reference frequency and voltage of VFDC converter, P_0 , Q_0 are the rated active and reactive power, P and Q are the actual active and reactive power output, m_p and n_q are the droop coefficient of active and reactive power for VFDC, ω_0 , v_0 are the nominal frequency and voltage respectively.

By finding the ω_{ref} , v_{ref} the VFDC based converter control can be implemented with the virtual impedance loop, voltage loop and current loop shown in fig.6.

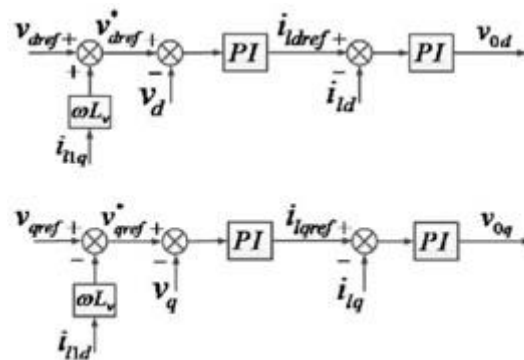


Fig. 6. Virtual impedance loop, voltage loop, and current loop of VFDC ($i_{id} * \omega L_v$ and $i_{iq} * \omega L_v$ denote the virtual impedance loop, respectively).

C. PQDC Control

The PQDC-based ESS controls certain active power and reactive power based on the system frequency and local voltage changes from their rated values.

Same as in VFDC control, in the droop control loop with the help of virtual inductor voltage v_v , the reference voltage v^* is corrected to compensate the drop in virtual inductor L_v as shown in fig.7.

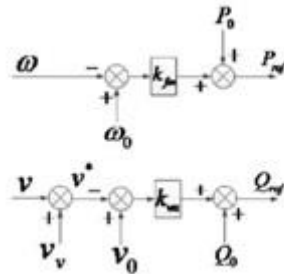


Fig. 6. PQDC control loop with voltage correction

$$\begin{aligned}P_{ref} &= P_0 + k_{fm}(\omega_0 - \omega) \\ Q_{ref} &= Q_0 + k_{vn}(v_n - v^*)\end{aligned}\quad (18)$$

where k_{fm} and k_{vn} are the droop coefficient of frequency and voltage for PQDC.

By finding P_{ref} and Q_{ref} , the PQDC converter control can be implemented with the virtual impedance loop, voltage loop and current loop as shown in fig.8 and 9 respectively.

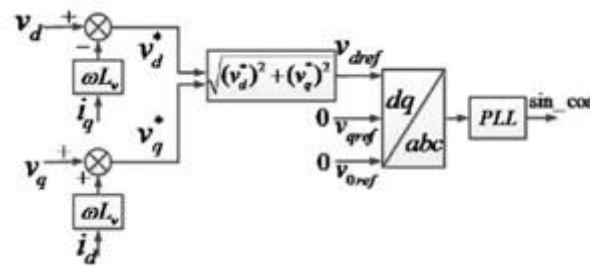


Fig. 8. Virtual impedance loop of PQDC.

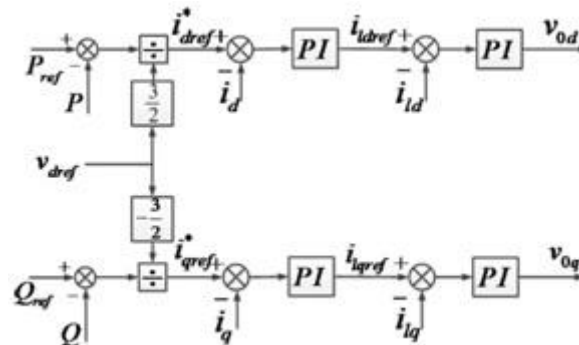


Fig. 9. Voltage and current loop of PQDC.

IV. Simulation Results and Discussion

In this paper autonomous microgrid performance is evaluated in terms of drooping characteristics of voltage with extensive MATLAB Simulations are carried out by varying load parameters.

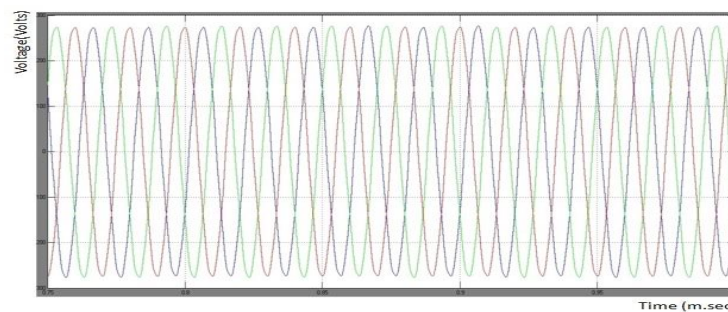


Fig10: Voltage characteristics of MG at 1000KW resistive load, 100KVAR inductive load, 50KVAR capacitive load

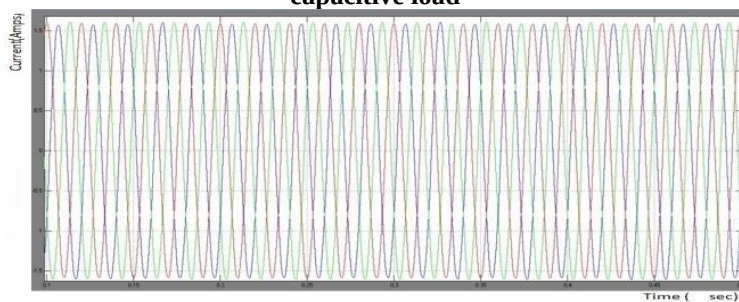


Fig11: Current characteristics of MG at 1000KW resistive load, 100KVAR inductive load, 50KVAR capacitive load

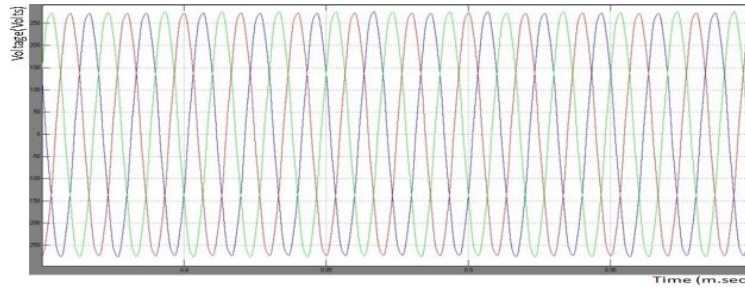


Fig12: Voltage characteristics of MG at 1200KW resistive load, 150KVAR inductive load, 80KVAR capacitive load

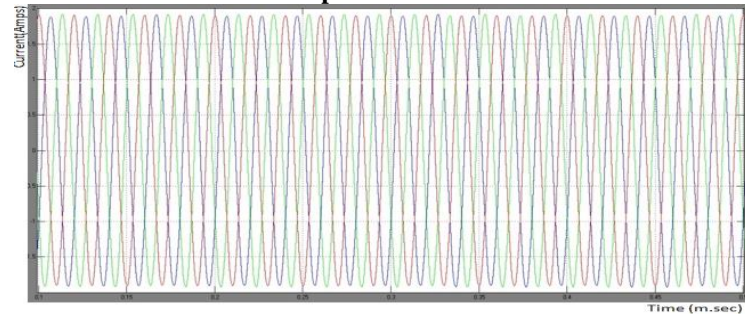


Fig13: Current characteristics of MG at 1200KW resistive load, 150KVAR inductive load, 80KVAR capacitive load

Table1: Voltage and frequency of MG at various loads

P	Q _L	Q _C	V	I	f	T.H.D
800	80	40	276	1.3	50	1.77
1000	100	50	275	1.6	50	1.85
1200	150	80	274	1.75	50	1.85
1500	200	100	272	2.5	50	1.9

V. CONCLUSION

In This paper a hybrid VFDC/PQDC droop control scheme for ESS-based autonomous MG. The proposed control structure VFDC ESS controls voltage angle and amplitude based on the active and reactive power, and PQ droop controlled ESS controls the active and reactive power output based on frequency and voltage deviations from the nominal values. The hybrid VFDC/PQDC can prevent the interference of line impedance uncertainty. The basic operation principle and control scheme of ESSs-based MG with hybrid VFDC/PQDC control is analyzed in detail.

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